# Binary Frequency Shift Keying (BFSK): Signal Representation

In binary FSK, the frequency of the carrier signal is varied to represent the binary digits and 0 by two distinct frequencies. The amplitude and frequency remain constant during each bit interval.

#### Signal Representation (coherent FSK)

Send:  $s_1(t) = Acos(2\pi(f_c + \Delta f))t)$  if the information bit is "1";

Send:  $s_2(t) = Acos(2\pi(f_c - \Delta f))t)$  if the information bit is "0";

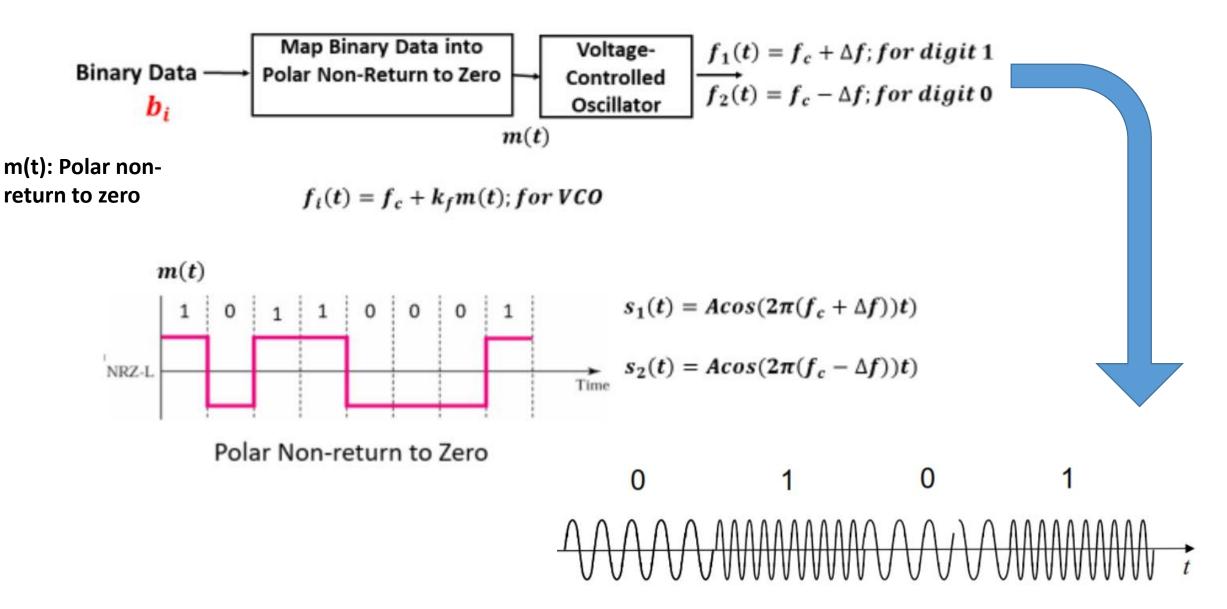
 $\Delta f$  is an offset frequency (from the unmodulated carrier  $f_c$ ) chosen so that  $s_1(t)$  and  $s_2(t)$  are orthogonal, i.e.,

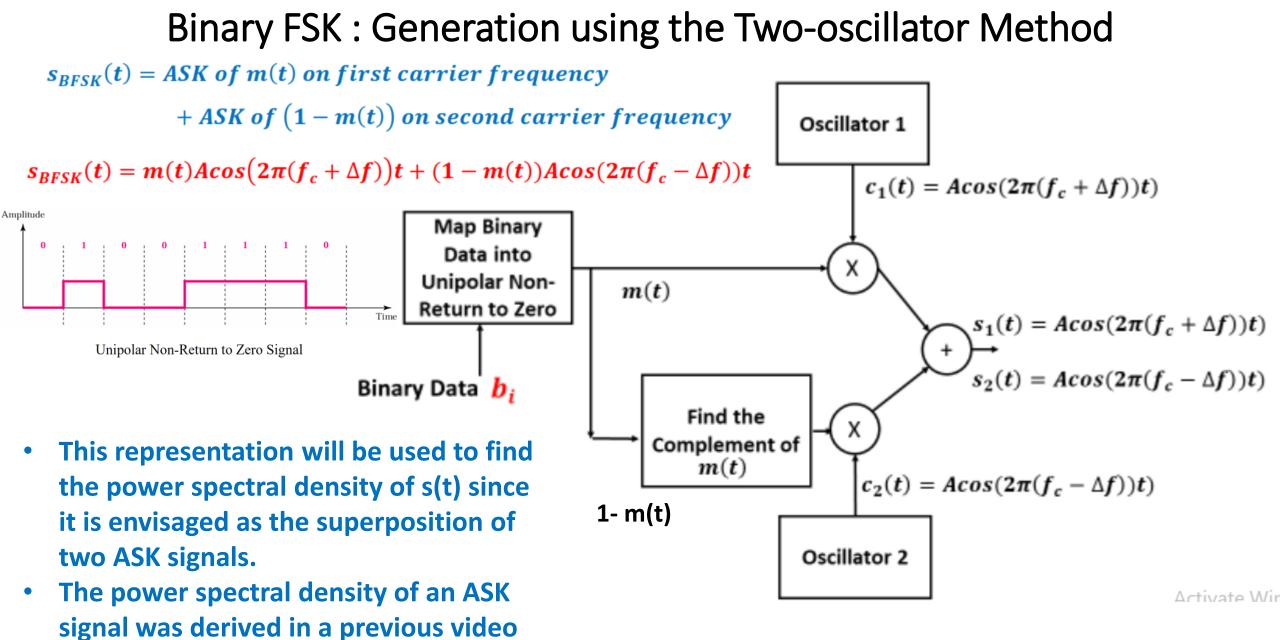
$$\int_0^{\tau} s_1(t) s_2(t) dt = 0$$

$$\frac{\sin(2\pi(2f_c)\tau)}{2f_c} + \frac{\sin(2\pi(2\Delta f)\tau)}{2\Delta f} = 0$$
Activate

Binary Frequency Shift Keying (BFSK): Signal Representation  $2f_c = \frac{n}{2\tau} = \frac{nR_b}{2}, n = 1, 2, \dots, f_c = \frac{nR_b}{4} = kR_b$ **Orthogonality condition**  $\frac{\sin(2\pi(2f_c)\tau)}{2f_c} + \frac{\sin(2\pi(2\Delta f)\tau)}{2\Delta f} = 0$  $2\Delta f = \frac{mR_b}{2}, m = 1, 2, \dots$   $\Delta f = \frac{mR_b}{4}$ Note that sin(x) = 0 The minimum frequency separation  $2\Delta f = R_b/2$ . when  $x = n\pi$ "1" "೧"  $\tau$ : is the time allocated  $\leftarrow \tau \rightarrow$ to transmit the binary digit.  $s_1(t) = A\cos(2\pi(f_c + \Delta f)t) \quad s_2(t) = A\cos(2\pi(f_c - \Delta f)t)$  $T_c = 1/f_c$  is the carrier period  $0 \leq t \leq \tau$  $0 \leq t \leq \tau$  $R_b = \frac{1}{\tau}$ : Data rate bits/sec  $(\tau \text{ is an integer number of } 1/(f_c \pm \Delta f))$ 

# Binary FSK : Generation using the Single Oscillator Method



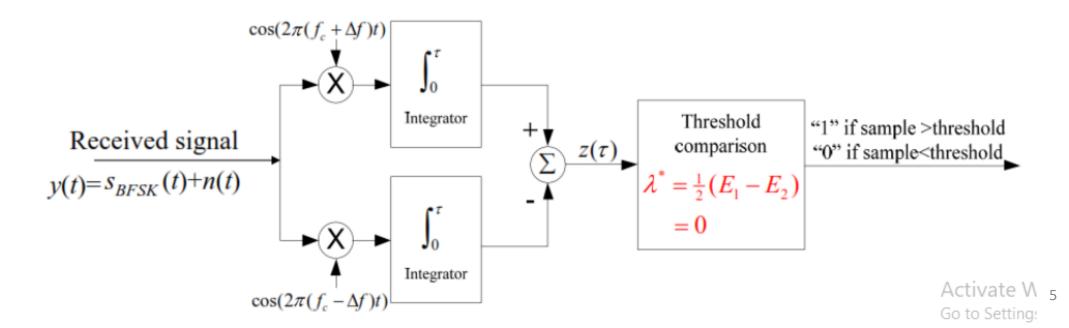


titled: Binary ASK

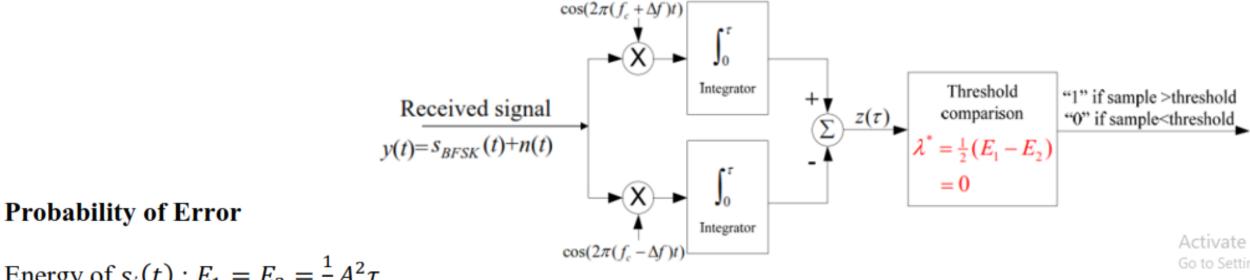
FSK: modeled as a sum of two ASK signals

### Binary FSK : Coherent Demodulation

The optimum coherent receiver consists of two correlators. The operation of the receiver makes use of the orthogonality condition imposed on the signals  $s_1(t)$  and  $s_2(t)$ . In the absence of noise, if  $s_1(t)$  is received, then the ouput of the upper correlator will have a value greater than zero, while the output of the lower correlator is zero. The converse is true when  $s_2(t)$  is received. In the presence of noise, the system decides 1 when  $z(\tau) >$ 0. That is, when the output of the upper correlator is greater than the output of the lower one. Otherwise, it decides 0.



# Binary FSK : Probability of Error



Energy of  $s_i(t) : E_1 = E_2 = \frac{1}{2}A^2\tau$ 

Average Energy per bit:  $E_b = \frac{1}{2}(E_1 + E_2) = \frac{1}{2}A^2\tau$ 

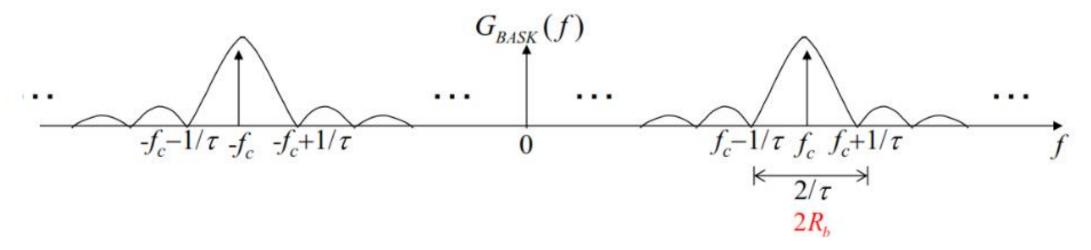
When the signals are orthogonal, i.e., when  $\int_0^{\tau} s_1(t)s_2(t)dt = 0$ , the probability of error is given by

$$P_b^* = Q\left(\sqrt{\frac{A^2\tau}{2N_0}}\right) = Q\left(\sqrt{\frac{E_b}{N_0}}\right)$$

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## Binary FSK : Power Spectral Density

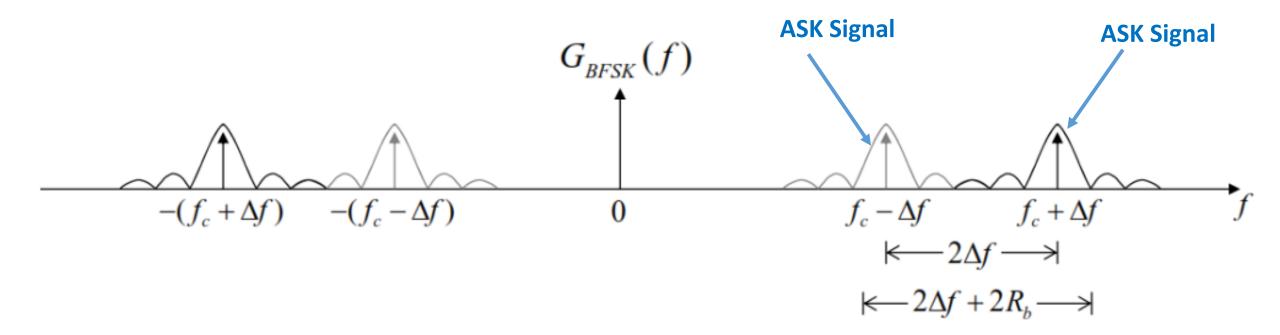
Since the FSK signal is the superposition of two ASK signals on two orthogonal frequencies, the spectrum is also the superposition of that of the ASK signals. We recall that the spectrum of the ASK signal is as shown below



 $s_{BFSK}(t) = ASK \text{ of } m(t) \text{ on first carrier frequency}$ + ASK of (1 - m(t)) on second carrier frequency

 $s_{BFSK}(t) = m(t)Acos(2\pi(f_c + \Delta f))t + (1 - m(t))Acos(2\pi(f_c - \Delta f))t$ 

### Binary FSK : Power Spectral Density and Bandwidth



The required channel bandwidth for 90% in-band power

$$B_{h_{90\%}} = 2\Delta f + 2R_{b}$$
  
B. W = (f\_1 - f\_2) + 2R\_b =  $\frac{R_b}{2} + 2R_b$ 

### **Binary FSK : Non-coherent Demodulation**

